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DSP-Based Phase Locked Loop Design for Three Phase Grid Synchronization

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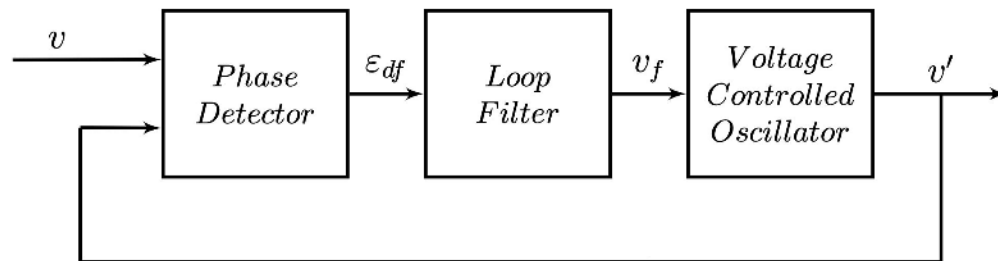
I. Introduction

Is accurate phase angle estimation methods needed for active and reactive power flow (Power Factor) applications?



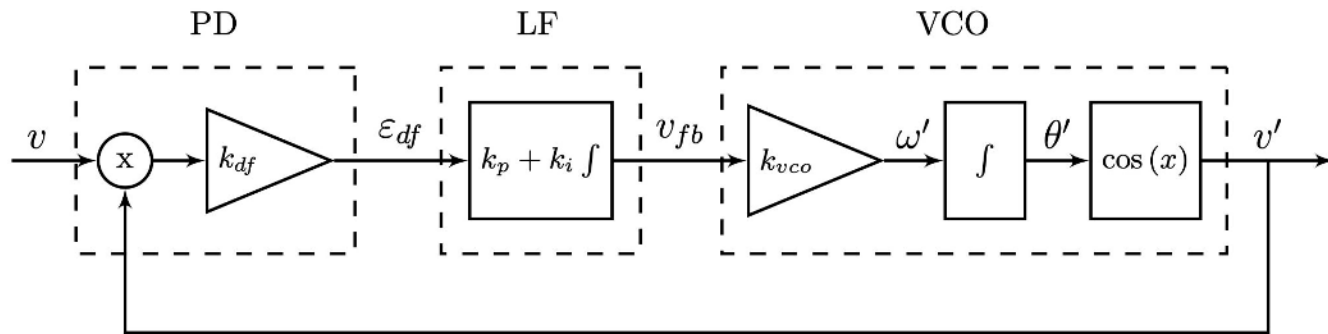
II. Theoretical aspects

Generated a output signal proportional to phase difference (error) and a characteristic of low pass filter for attenuate the components of high frequency and generated the output the same frequency that the reference signal



III. Proposed methodology

Subtitulo



Phase detector

$$E_{df}(s) = \frac{1}{2} (\Theta(s) - \Theta'(s))$$

Loop filter

$$V_{fb}(s) = \left(k_p + \frac{k_i}{T_i s} \right) E_{df}(s)$$

Voltage Controlled Oscillator

$$\Theta'(s) = \frac{1}{s} V_{fb}(s)$$

III. Proposed methodology

Subtitulo

$$LF(z) = \frac{\left(\frac{2k_p + k_i * T}{2}\right) - \left(\frac{2K_p - K_i * T}{2}\right) z^{-1}}{1 - z^{-1}}$$

Based on the discrete function of the low-pass filter block, we proceed to show the DSP implementation of each of the phase tracking loop blocks in pseudocode.

```
void SPLL_lph_Control (SPLL_lph *spll)
{
    // Phase Detect
    spll->PD=(spll->AC_input*spll->cos);

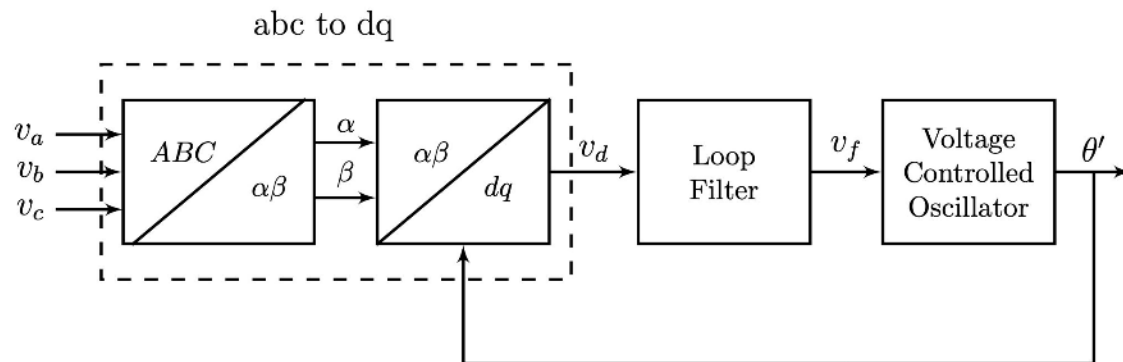
    // PI loop filter
    spll->y_lf=
        -(spll->lpf.K2*spll->y_lf)
        +(spll->lpf.K0_lf*spll->y_pd)
        +(spll->lpf.K1_lf*spll->y_pd);

    // VCO
    spll->wo=spll->wn+spll->y_lf;
    spll->theta=spll->theta_old
        +(spll->wo*spll->T);
    spll->cos=cos(spll->theta)

    // Update for next iteration
    spll->y_lf_old=spll->y_lf;
    spll->theta_old=spll->theta;
}
```

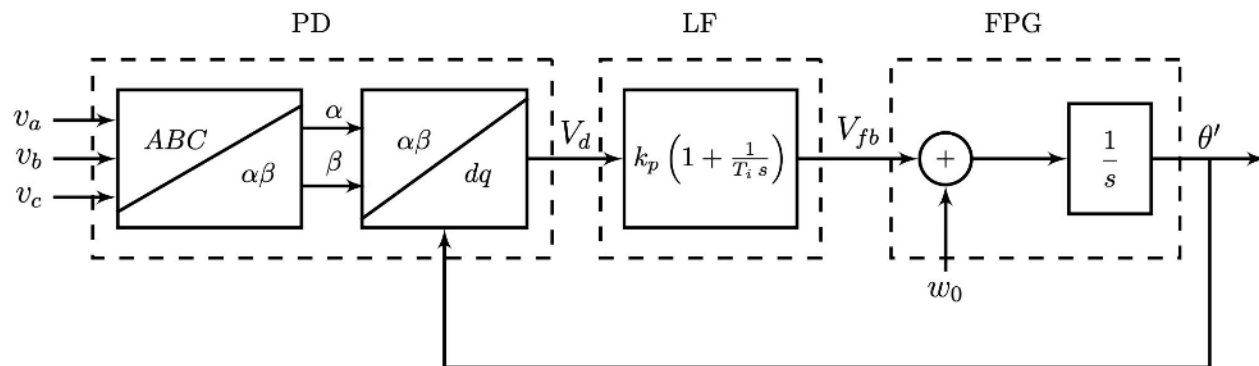
II. Theoretical aspects

Common methods for synchronizing networks proposed in the literature are, open loop model system such as filters, but these model system do not work correctly when the system is unbalanced in voltage source (Voltage).



III. Proposed methodology

Subtítulo



Alpha - Beta transform

$$\begin{bmatrix} x_\alpha \\ x_\beta \end{bmatrix} = \frac{2}{3} \begin{bmatrix} 1 & -\frac{1}{2} & -\frac{1}{2} \\ 0 & -\frac{\sqrt{3}}{2} & \frac{\sqrt{3}}{2} \end{bmatrix} \begin{bmatrix} x_a \\ x_b \\ x_c \end{bmatrix}$$

Quadrature - Zero transformation

$$\begin{bmatrix} x_d \\ x_q \end{bmatrix} = \begin{bmatrix} \cos(\theta) & \sin(\theta) \\ -\sin(\theta) & \cos(\theta) \end{bmatrix} \begin{bmatrix} x_\alpha \\ x_\beta \end{bmatrix}$$

III. Proposed methodology

Subtitulo

```
void ABC_DQ0 (ABC_DQ0 *DQ)
{
    //Reverse Transform - Alpha Beta
    DQ->alpha=(2/3) * (DQ->a-(1/2)
                    * (DQ->b+DQ->c) );
    DQ->beta=(2*sqrt(3)/6) * (DQ->b-DQ->c);
    //Park Transform - DQ
    DQ->z=(1/3) * (DQ->a+DQ->b+DQ->c);
    DQ->d=DQ->alpha*DQ->sin
        -DQ->beta*DQ->cos;
    DQ->q=DQ->alpha*DQ->cos
        +DQ->beta*DQ->sin;
}
```

The values of the three phases are transformed into the rotating reference frame and the q component is used as phase detection

III. Proposed methodology

Subtitulo

This results in the high-frequency term of the error signal not being taken into account when studying the dynamic response of the phase-locked loop.

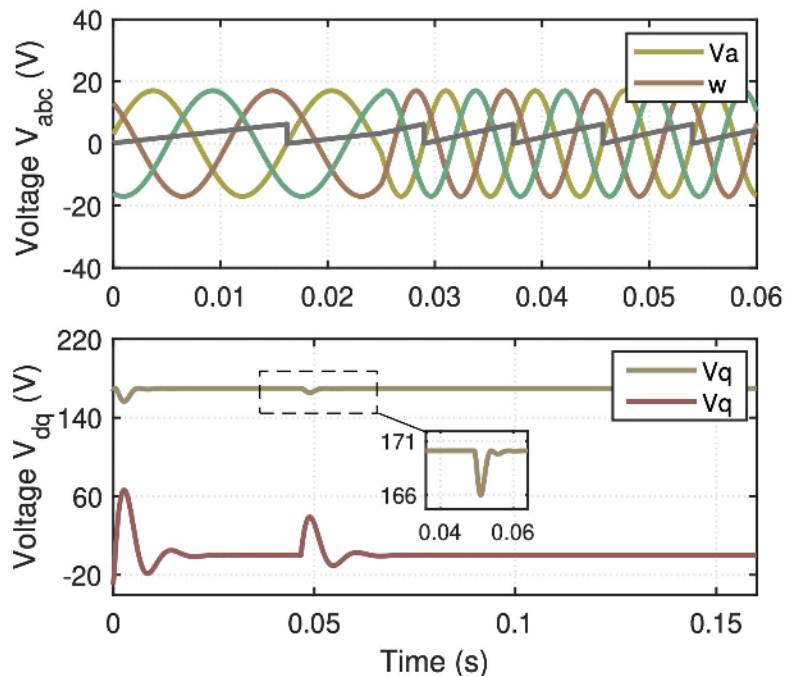
```
void SRFPLL_Control(SFR_PLL *sp11)
{
    // Loop Filter
    sp11->y1f=sp11->y1f_old
        + (sp11->lpf.K0*sp11->v_q)
        + (sp11->lpf.B1*sp11->v_q_old);

    // VCO
    sp11->fo=sp11->fn + sp11->y1f;
    sp11->theta=sp11->theta_old
        + (sp11->fo*sp11->T) * (2*pi);

    if(sp11->theta[0] > 2*pi)
        sp11->theta=sp11->theta - (2*pi);

    // Update for next iteration
    sp11->y1f_old=sp11->y1f;
    sp11->v_q_old=sp11->v_q;
    sp11->theta_old=sp11->theta;
}
```

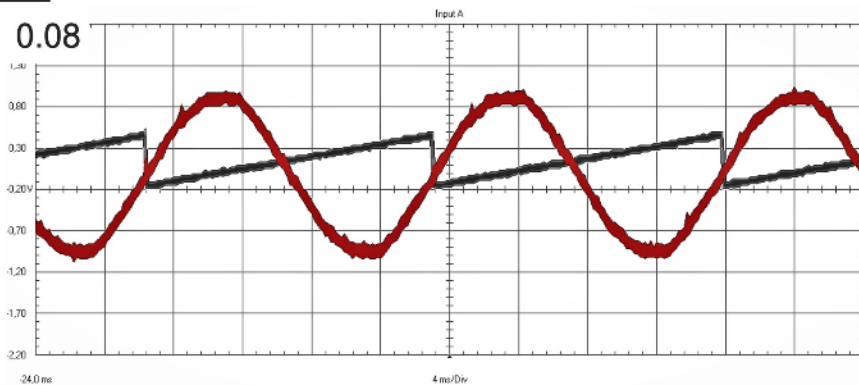
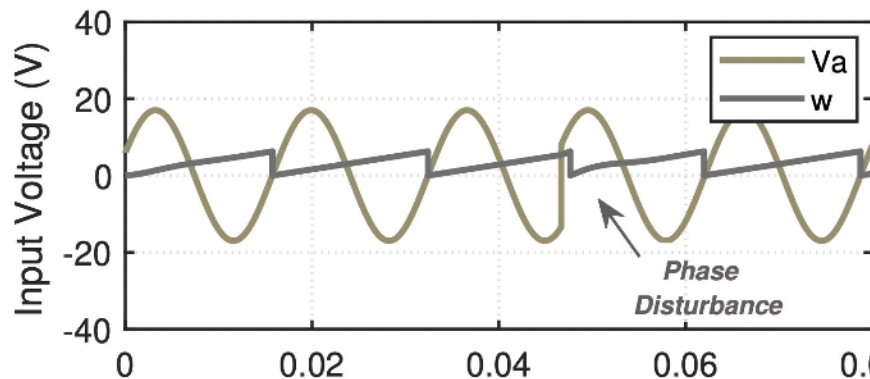
IV. Results



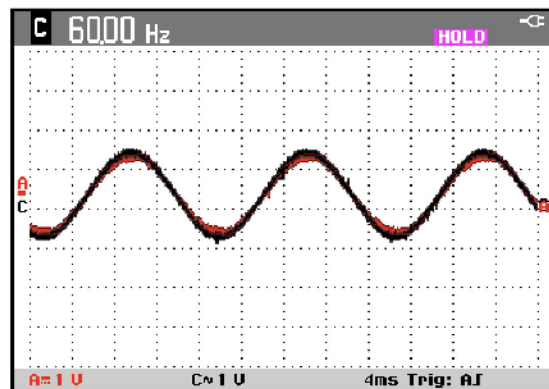
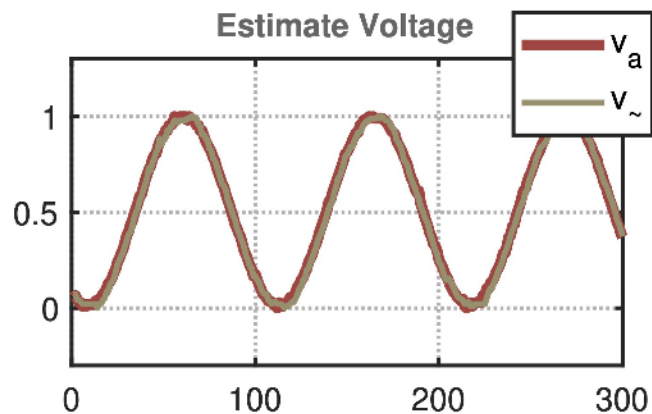
When using Synchronous Reference Framework PLL (SFR PLL), Replaced by the synchronous reference framework PLL they he transform three-phase values `abc' that are variable in time, to constant values in permanent regime

IV. Results

Phase Locked Loop



IV. Results

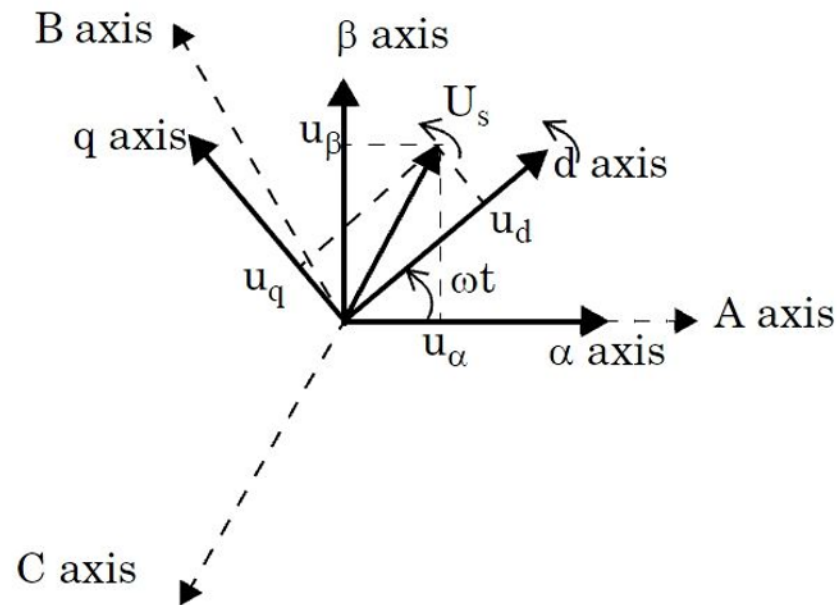


$$RMSE = \sqrt{\frac{1}{n} \sum_{i=1}^n (y_i - \hat{y}_i)^2}$$

The data of the estimation and measurement of the line voltage is obtained and a normalization of the same is performed to be compared with the estimation

V. Conclusions

The model system can be implemented in different voltage conditions of input and the simplicity of this structure makes to the system of phase-locked presented be appropriate for the digital implementation.



V. Conclusions



Equally the correct functioning of phase-locked loop with disturbances, demonstrating better yield of estimate of angle respect to other methods propose in the literature and this method follow frequency variations

QUESTIONS ?

VII. References

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